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Research paper

## Linear array Yagi-Uda 5G antenna for vehicular application

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#### **Abstract**

A two element 2X2 linear array Yagi antenna which is meant to operate at frequency of  $3.5 \, \text{GHz}$  which can be used for low band  $5 \, \text{G}$  as well as for WIMAX is presented in this paper. This array consists of two radiating elements of eight parasitic elements each with overall dimensions of  $110 \times 60 \times 1.6 \, \text{mm}$ 3. The proposed antenna was built on Rogers Duroid substrate. It exhibits a good bandwidth (impedance bandwidth of single antenna element about  $600 \, \text{Mhz}$ .for S11 less than- $10 \, \text{dB}$  at the center frequency of  $3.5 \, \text{GHz}$ ) and gain of  $10.5 \, \text{dB}$ . The idea behind this paper is to attain good gain along with bandwidth which can be utilized for vehicular applications.

### 1. Introduction

One of the most convenient and luxury style of transport is by car, so it has become necessary to implement all types of luxurious entertainment, driver information and communication systems in a vehicle. The next generation of wireless communication which is Fifth generation or 5G holds a high rate of adoption in automobile industry as it provides better communication and different unique applications like multi person video calling, autonomous driving, real time gaming and many more on go, which present 3G or 4G cannot provide. Because of its low latency and high speed it can be enabled in vehicles to communicate with outside world and nearby vehicles which could result in significantly more efficient and well organized traffic system. If all of the vehicles on a road and road side objects were connected to form a network, they can share safety information as a part of incorporating a Intelligent traffic management system. Hence Vehicle to vehicle communications can pave way for potentially travel at much higher speeds and within greater proximity of each other without risk of accident. Due to favourable properties the frequency bands for 5G applications ranges for lower 5G bands below 6GHz is 3.4-3.8 GHz and 4.4 to 4.9 GHz are deployed for vehicular applications. It has become utmost important to design an antenna with small size, high gain and broad beam-width coverage for this vehicular communications. Its well known that in antennas used for communications suffer from most common problem which is path loss. To prevail over this problem adopting an antenna array is much suitable solution to increase gain as well as physical aperture.

The first fabricated microstrip Yagi antenna was introduced by Huang [4], which comprises one reflector element, one driven element, and two director elements. That microstrip Yagi antenna was able to achieve a gain of around 8dB. A typical microstrip Yagi antenna consists of a driven patch element and a few parasitically coupled reflector and director patch elements [5]. It is a well-established fact that the directivity of microstrip Yagi antenna greatly depends on the number of directors [6]. Various at

tempts have been made to enhance the gain of Yagi antenna by incorporating more than a single director [7–12].

## 1.1 Introduction to proposed antenna

The conventional Yagi antenna is a directional antenna which is well known for its high directivity and gain but the drawback of it was its size and it is not conformal because of its structure. On the other hand the microstrip antenna which is small in size, light weight and compatible with other devices paves way for fabrication of microstrip quasi Yagi antenna where Yagi antenna like structure is designed on a microstrip antenna which possesses the features of high gain and directivity as the conventional Yagi antenna. it also has the advantages of wide bandwidth, low cost and easy to fabricate.

#### 1.2 Mechanism of Microstrip Yagi-Uda

In a Microstrip Yagi –Uda the electromagnetic energy is coupled from the driven element dipole through space into the parasitic dipoles and then gets reradiated to form a directional beam. This mechanism is analogous to conventional Yagi antenna, but the slight difference is that the substrate also plays some role in coupling by surface waves. The microstrip patch usually couples very little energy as it radiates largely in broadside and slightly along the direction of the ground plane [13-17]. To attain maximum behavior like a conventional Yagi-Uda antenna the microstrip patches has to place vey near, the separation should be less in between parasitic patches to attain significant amount of coupling. To attain maximum coupling the spacing between the elements should be equal to or less than the thickness of the substrate and this is experimentally found.[18-20].

In this paper we have proposed a microstrip Yagi antenna for 5G applications in a vehicle because of its high directivity and high gain which is suitable for wireless communications. The proposed antenna is a two element linear array which can obtain a high gain and good bandwidth. The simulated results shows the gain and reflection coefficients attained by antenna.

# 2. Design of single element microstrip Yagi antenna

The Geometry and detailed dimensions of the proposed antenna element is shown in Fig.1. The proposed Quasi-Yagi antenna is printed on the Rogers Duroid 5880 substrate with dielectric constant of 2.2, conductor loss (tan  $\delta$ ) of 0.0009 and thickness of 1.6 mm.. The antenna is matched with 50 $\Omega$  input impedance. The antenna consists of three main components, a dipole driver, the parasitic directors and the reflector element. The ground plane acts as a reflector to achieve a directive radiation pattern. To further improve the directivity, eight parasitic directors are designed and optimized at the operating frequency of 3.5GHz.

As per design equations of the yagi antenna the length of the driver is  $0.5\lambda_g$  and the length of the director is  $0.45\lambda_g$  where ' $\lambda_g$ ' is the effective wavelength of the antenna and can calculate by

$$\lambda_g = \frac{c}{fr\sqrt{\varepsilon_r + 1}} \tag{1}$$

Where C is the velocity of the Electromagnetic wave of free space,  $\mathcal{E}_r$  is the relative dielectric constant of the substrate,  $f_r$  is the resonant frequency of the antenna. We can calculate the width of the radiators by formula

$$w = \frac{c}{fr\sqrt{2(\varepsilon_r + 1)}}\tag{2}$$

We can calculate the value of each parameters by formula (1) and (2). Which were shown in table 1.

Table 1: Antenna Dimensions in mm

L	W	L1	W1	Wf	W2	W3	W4
55	60	20.8	11.662	2.7	17.975	3.28	1

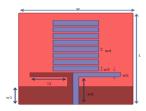


Fig. 1: Proposed Yagi-Uda Antenna Geometry

In this design, the antenna simulations are done using the full-wave electromagnetic field simulator Ansoft HFSS version 13. The Fig.2, depicts the simulated return loss (S11) of the proposed single microstrip Quasi-Yagi antenna. The operating bandwidth is about 620MHz for S11 less than - 10dB at the center frequency of 3.5GHz of the antenna can be suitable for array design. The Fig 5, depicts the far field radiation pattern of E and H plane of a single antenna element. The antenna peak gain is more than 7dB. The above single element Yagi antenna yields a Far field gain of more than 7dB but in order to acquire more gain from this model the next step of modification is implementing a linear array for the above model. Here we have proposed the 2 element linear array Yagi antenna which can attain better gain, bandwidth and radiation pattern compared to single element.

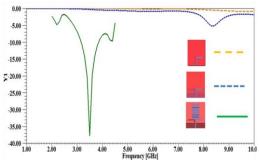


Fig. 2: Step by Step Construction of Yagi Antenna and corresponding S11

## 3. Proposed 2 Element Linear Array Yagi-Uda Antenna

The schematic of the 2 element linear array of the microstrip Quasi-Yagi antenna is shown in Fig.3. In this design,  $1\times2$  uniform linear array (sub-array) antennas have been used, where each radiating element of them is excited by signals with equal magnitude. The simulated S-parameters of the proposed array antenna structure is illustrated in Fig.4. And fig.4 and fig.5 shows the Reflection coefficient and radiation pattern of the proposed antenna Yagi-Uda antenna respectively.

As shown in Fig.3, all the 16 parasitic elements perform very consistently with better S11 and acceptable bandwidth to cover 3.5 GHz

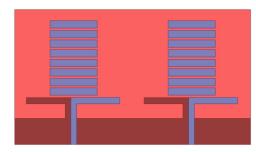


Fig. 3: Two Elements Linear Array Yagi-Uda Antenna

## 4. Results and Discussions

## **4.1 Reflection coefficient and Current Distribution of Single Antenna**

The reflection coefficient for the eight elements single Yagi antenna is as shown below. The figure 4 depicts the S11 value of -36.2 with a band width of 500 Mhz.

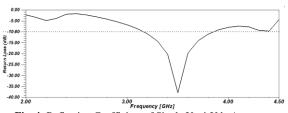


Fig. 4: Reflection Coefficient of Single Yagi-Uda Antenna

## 4.2 Radiation pattern of single 8 element Yagi antenna

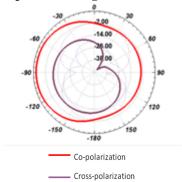


Fig. 5: Radiation Pattern of Yagi-Uda Antenna

It is observed from the figure 5 that the co polarization is more compared to cross polarization which is desirable quality of working of good antenna.

## 4.3 Current Distribution of two element linear array

In The Yagi-Uda array the surface travelling wave is described by the current distribution in each element and by the phase velocity of the travelling wave. In calculation of current distribution in the Yagi-Uda antenna where the currents are induced in the parasitic elements via the mutual impedances with the exciter, and radiation occurs as from a set of discrete sources. Green[10] has analyzed arrays with up to ten elements using this method with good results.

Current distribution in two element linear array at 3.5 GHz is 6.4 A/m.

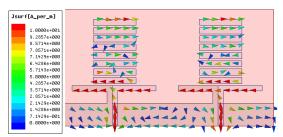


Fig. 6: Current Distribution in Two Element Linear Array

## 4.4 Envelope Correlation Coefficient (ECC)

Envelope Correlation Coefficient tells us how independent two antennas' radiation patterns are. So if one antenna was completely horizontally polarized, and the other was completely vertically polarized, the two antennas would have a correlation of zero. Similarly, if one antenna only radiated energy towards the sky, and the other only radiated energy towards the ground, these antennas would also have an ECC of 0. Hence, Envelope Correlation Coefficient takes into account the antennas' radiation pattern shape, polarization, and even the relative phase of the fields between the two antennas. The antenna structures in this study exhibit an envelope correlation coefficient of 0 at 3.5 GHz as shown in Figure 7. And also it is below 0.5 for the entire bandwidth of the proposed antenna which shows better isolation between two antenna elements in array.

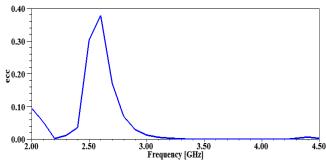


Fig. 7: Envelop Correlation Coefficient of Array Antenna

## 5. Parametric Analysis

All significant physical parameters, such as W5,W4,W3,L1,substrate materials should be adjusted carefully in order to achieve a good performance. In this section, the effects of these parameters on impedance bandwidth and realized gain are examined to facilitate the design of a same or homogeneous antenna with different specification.

## 5.1 Different lengths of dipole director

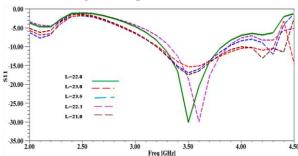


Fig. 8: Parametric analysis for different lengths of dipole director

The antenna which is modeled in the paper needs significance to optimize the length of the dipole driver, width of the driver, space between parasitic directors and substrate. The S parameters of the length of drivers were shown in figure8. Each value of the length is 23.8mm, 24.3mm, 22.8mm, 22.3, and 21.8mm. For length 23.8mm and S11 at 3.5GHz is -15.3865, realized gain is 8.23 and bandwidth is 800Mhz.

For length 24.3mm and S11 at 3.5GHz is -16.9845,realizedgain is 8.65 and bandwidth is 600Mhz. For length 22.3mm and S11 at 3.5GHz is -17.48,realized gain is 8.59 and bandwidth is 800Mhz.For length 22.8mm and S11 at 3.5GHz is -26.67,realized gain is 10.5 and bandwidth is 620Mhz.For length 21.8mm and S11 at 3.5GHz is -30.10,realized gain is 8.37 and bandwidth is 500Mhz.

## 5.2 Different widths of director dipole

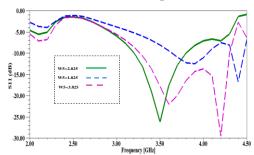


Fig. 9: Parametric analysis for different widths of director dipole

The S parameters of the width of directors were shown in figure 9. Each value of the length is 2.825,3.825mm and 1.825mm. From

the simulation curve in figure 5, when the length is 3.825mm, the error of the curve is bigger.

#### 5.3 Parametric analysis of different substrate

The changes in value of S11 parameters using different substrate materials were shown in figure 10.

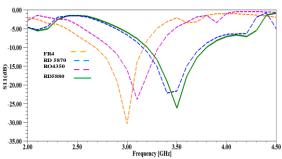


Fig. 10: Paramentric analysis for different substrate material

Each substrate exhibits its own properties by which S parameters, bandwidth and gain of the antenna changes accordingly. S11 of Rogers Duroid 5880 is found to be -26.18 with bandwidth of 600Mhz,S11 of Rogers Druroid 5870 is -21.046 with bandwidth of 500Mhz,S11 of Rogers 4350 is -24.472 bandwidth 400Mhz ,S11 of Fr4 -32.24 with bandwidth of 390Mhz. From above analysis it is obvious that RT Duroid 5880 is providing greater bandwidth comparatively.

### 5.4 Different spaces between parasitic patches

The S parameters are optimized with different values but equal gap between parasitic patches of drivers were shown in figure 11.

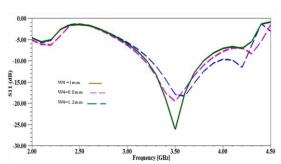


Fig. 11: Parametric analysis for different spaces between parasitic patches

Each value of the length 0.8mm,1mm and 1.2mm. The S parameters of the space between the parasitic element are shown in figure 11. From the simulation curve in figure, when the length is 1.2mm the error of the curve is bigger. The curve near center frequency with the distance s=1mm.

When width w4 is 1.2mm then gain of antenna is 7.7 and bandwidth is 600Mhz. When width w4 is 0.8mm then gain of the antenna is 8.61 and bandwidth is 700Mhz.

Table 2	: Compari:	son with	Previous	Work

Table 2: Comparison with Previous Work							
Work	Antenna	actua-	Bandwidth	Gain	Frequency		
	Size	tors					
[1]	136×68	MIM O	200Mhz	NR	3.5GHz		
[11]	100×92.8	Yagi	160 Mhz	7.37d Bi	5.38GHz		
[ 13]	263 ×263	MIM O	400Mhz	7 dB	5.8GHz		
[5]	100X 80	Yagi	560Mhz	9.5d Bi	5.2GHz		
Pro- posed	110×60	Yagi	~600Mhz	10.6 dB	3.5GHz		

#### 6. Measures Result

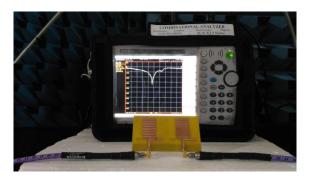


Fig 12. Measured data on VNA

#### 7. Conclusion

In this paper, we proposed a two element linear array antenna, which is composed of sixteen microstrip Quasi-Yagi parasitic elements and forms a 2×2 linear structure with center frequency operating in the 5G low-band at 3.5 GHz. The simulation results show that the antenna array has high gain (peak value about 10.5dB), wide bandwidth about 600MHz at the center frequency of 3.5GHz and can approximately achieved. Meanwhile, within the operating bandwidth, good F/B ratios 13.68 has been observed. With this we can conclude that this proposed antenna has acquired almost the desirable characteristics which suits better for the vehicular antenna and fits better in 3.5GHz low band of 5G.

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